

# Improved Split CV Mobility Extraction in 28 nm Fully Depleted Silicon on Insulator Transistors

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**Abstract**—In this work, we assess the applicability of the well-known split CV technique for mobility extraction in 28 nm FDSOI transistors with gate length down to 25 nm using TCAD simulations. We identify the significant bias dependence of the parasitic source/drain resistance and the contribution of the inner fringing capacitance as the main sources of error in the conventional split CV extraction. An improved split CV method, correcting for these parasitics, is demonstrated to accurately extract the effective mobility and its dependence on the gate voltage for devices down to 25 nm gate length. Measurements on 28 FDSOI transistors confirm the insights from the TCAD simulations.

**Index Terms**—FDSOI, UTBB, split CV, effective mobility.

## I. INTRODUCTION

EXTRACTION of the effective mobility  $\mu_{eff}$  is of major importance to evaluate the performances of advanced transistors. This has become challenging in highly scaled nodes due to the increased relative importance of parasitics, thus requiring adaptations of the well-known methods to properly correct for them. The split current-voltage (CV) method [1] has regained attention for its robustness compared to other simple methods, and several modifications have been proposed to correct it for gate leakage, parasitic capacitances and access resistances [2]–[8]. In this letter, we apply the split CV method to advanced FDSOI devices by identifying and rigorously de-embedding the different parasitics affecting the extraction. We confirm and quantify their impact on the obtained  $\mu_{eff}$  using TCAD simulations and propose an adapted method to correct for them, not requiring any pre-assumed mobility or inversion charge voltage dependence. It is validated experimentally on 28 FDSOI devices down to 25 nm gate length.

## II. EXTRACTION PROCEDURE

The split CV method relies on the expression of the drain current  $I_D$  in strong inversion and linear regime, yielding [7]

$$\mu_{eff} = \frac{L_{eff}^2 I_D}{Q_i (V_{DS} - R_{SD} I_D)} \quad (1)$$

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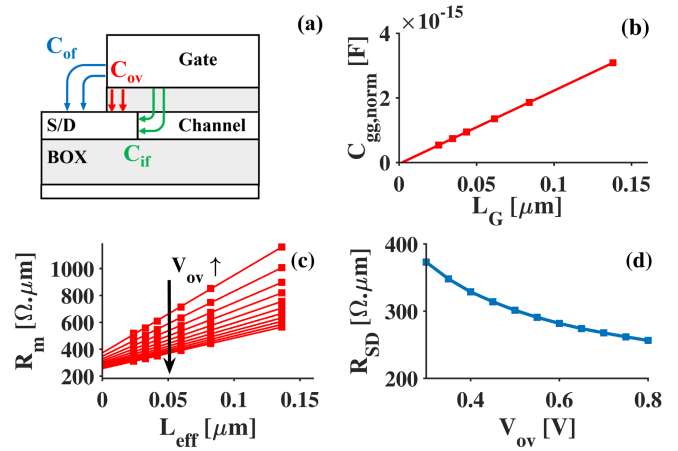
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**Fig. 1.** (a) Parasitic gate capacitances as modeled in [7] (b)  $L_{eff}$  extraction by linear interpolation of TCAD  $C_{gg, norm}(V_{ov, max} = 0.87 \text{ V}, V_{DS} = 0 \text{ V})$  curves for  $L_G = [25, 35, 45, 60, 90, 150] \text{ nm}$  (c)  $R_{SD}$  extraction with the total resistance method for 10 equally spaced voltages in  $V_{ov} = [0.3, 0.87] \text{ V}, V_{DS} = 50 \text{ mV}$  (d) Extracted  $R_{SD}(V_{ov})$  dependence.

The computation of  $\mu_{eff}$  therefore requires precise extraction of the effective length  $L_{eff}$ , the total series access resistance  $R_{SD}$  and the inversion charge  $Q_i$  as discussed in the following.

### A. Gate Capacitance and $L_{eff}$ Extraction

Figure 1a shows the parasitic capacitances usually considered for advanced MOSFET's. The total gate capacitance writes  $C_{gg}(V_{ov}, L_{eff}) = C_{inv}(V_{ov}, L_{eff}) + C_{ov} + C_{of}(L_{eff}) + C_{if}(V_{ov})$  where  $C_{inv}$  is the inversion capacitance,  $C_{ov}$ ,  $C_{of}$ ,  $C_{if}$  respectively the overlap, outer and inner fringing capacitances and  $V_{ov} = V_{GS} - V_{TH}$  the overdrive voltage [7]. In order to remove the extrinsic contributions, one classically defines  $C_{gg, norm}(V_{ov}) = C_{gg}(V_{ov}) - C_{gg}(V_{ov, min})$  with  $V_{ov, min}$  biasing the device in accumulation [7], [9].

In contrast to bulk devices, in FDSOI the inner fringing capacitance is not shielded in accumulation due to the absence of dynamic majority carrier response [10], [11]. One then has

$$C_{gg, norm}(V_{ov}, L_{eff}) \approx C_{inv}(V_{ov}, L_{eff}) + C_{if}(V_{ov}) - C_{if}(V_{ov, min}) \quad (2)$$

which shows a parasitic contribution of the voltage-dependent inner fringing capacitance. Extraction of  $L_{eff} = L_G - \Delta L$  is usually performed by interpolating  $C_{gg, norm}$  extracted in strong inversion against the gate length  $L_G$  [12], [13]. Though the inner fringing capacitance is known to be shielded in strong

inversion [10], [14], it has been noted that the  $C_{if}(V_{ov,min})$  term in Eq. (2) could disturb the extraction [11]. This will be assessed in Section III.

### B. Bias Dependent $R_{SD}$ Extraction

Correcting for the access resistance  $R_{SD}$  is a mandatory step in short transistors, where it can dominate the total device resistance. While several  $R_{SD}$  extraction methods are available in literature [15]–[17], they often neglect its bias dependence. However, in [17], [18] a significant dependence of  $R_{SD}$  on the gate voltage was reported in FDSOI and linked to the formation of a local inversion layer in the underlaps (i.e. beneath the spacers). This was shown to lead to an artificial degradation of the low-field mobility extracted using the Y function method [18]. Here,  $R_{SD}$  is therefore extracted for different overdrive voltages using the total resistance method [15], where  $R_{SD}$  is obtained as the linear interpolation at  $L_{eff} = 0$  of the total device resistance  $R_m(L_{eff}) = \frac{V_{DS}}{I_D}(L_{eff})$  curves (Fig. 1c).

### C. $Q_i$ Extraction Correcting for the Inner Fringing Capacitance

The inversion charge  $Q_i$  is ideally obtained as

$$Q_i(V_{ov}) = \int_{V_{ov}^*}^{V_{ov}} C_{inv}(V)dV \quad (3)$$

where  $V_{ov}^*$  is a bias voltage in accumulation (where  $C_{inv}$  is zero). Using  $C_{gg,norm}$  as an approximation of  $C_{inv}$  in Eq. (3) results in a parasitic charge contribution due to the integration of the inner fringing terms in Eq. (2). In this work, we therefore rather model the total gate capacitance  $C_{gg}$  as

$$C_{gg} \approx A(V_{ov})L_{eff} + B(V_{ov}) \quad (4)$$

with  $A(V_{ov}) = A_1(V_{ov}) + A_2$ ,  $B(V_{ov}) = B_1(V_{ov}) + B_2$  and perform a linear interpolation of  $C_{gg}$  against  $L_{eff}$  to extract these coefficients (Fig. 2b).  $A_1(V_{ov})$  can be attributed to the inversion capacitance and  $A_2$  to the outer fringing capacitance per unit length. The  $B(V_{ov})$  term corresponds to a usual parasitic capacitance plot extracted in a similar manner in [8], [17], [19].  $B_1(V_{ov})$  relates to the inner fringing behaviour, while  $B_2$  is finally attributed to the overlap capacitance.

## III. RESULTS AND DISCUSSION

In this section, Synopsys Sentaurus TCAD simulations of  $I_D(V_G)$  and  $C_{gg}(V_G)$  (obtained from the simulated  $Y_{11}$  parameter) curves are used to validate the extraction procedure and quantify the impact of the different parasitics on the extracted mobility. The bulk and source contacts are kept to ground. The 2D simulated structure was provided by CEA-Leti and adequately reproduces the 28 nm FDSOI transistor geometry and doping profiles [20].

A gate bias dependent mobility model (including effects of dopant impurity scattering, velocity saturation and surface scattering) was implemented in the simulation.

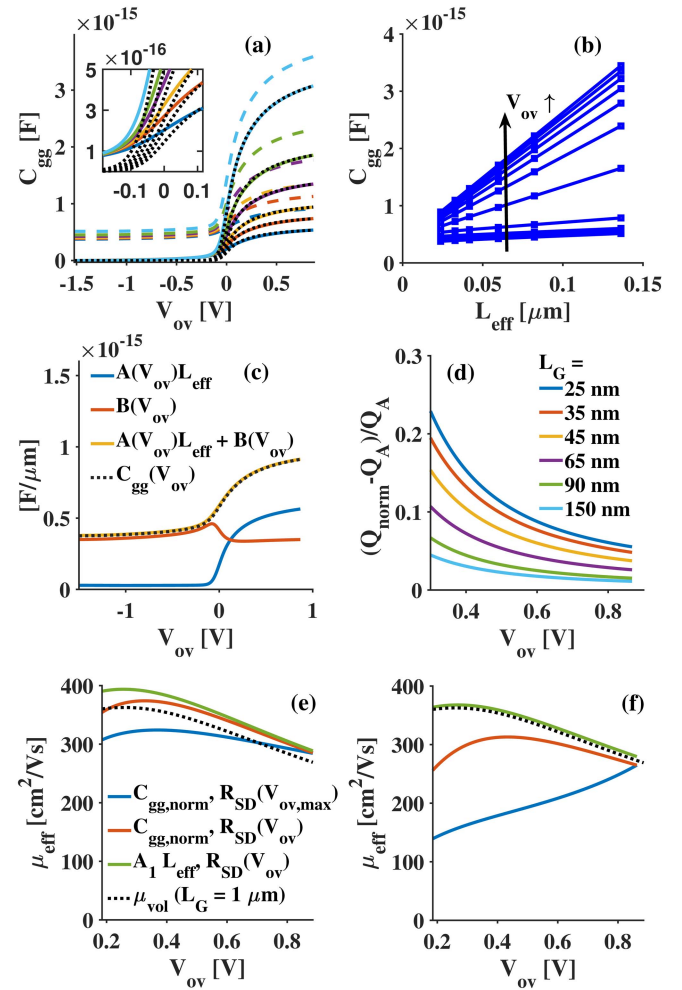


Fig. 2. (a)  $C_{gg}$  (dashed),  $C_{gg,norm}$  (full) and  $A_1L_{eff}$  (black dots) for  $L_G = [25, 35, 45, 60, 90, 150]$  nm (colors correspond to the legend in (d)),  $V_{DS} = 50$  mV,  $f = 100$  kHz. The inset highlights the difference between  $C_{gg,norm}$  and  $A_1L_{eff}$  due to the inner fringing terms. (b) Interpolation of  $C_{gg}$  against  $L_{eff}$  for 20 equally spaced overdrive voltages in  $V_{ov} = [-1.5, 0.87]$  V (c) Results from the linear interpolation of  $C_{gg}$  for a  $L_G = 25$  nm transistor (d) Relative error between the charge extracted from  $C_{gg,norm}$  ( $Q_{norm}$ ) and from  $A_1L_{eff}$  ( $Q_A$ ) for gate lengths between  $L_G = 25$  nm and  $L_G = 150$  nm (e),(f)  $\mu_{eff}$  extraction for transistors with resp.  $L_G = 150$  nm,  $L_G = 25$  nm ( $V_{ov,max} = 0.87$  V). The effective mobility implemented in TCAD is estimated as  $\mu_{vol} = \frac{\iint_S \mu(x,y)n(x,y)dx dy}{\iint_S n(x,y)dx dy}$  where  $S$  is the effective channel region. A long  $L_G = 1 \mu\text{m}$  transistor has been used to minimize the influence of the source and drain doping diffusions (corrected by  $R_{SD}$ ) in the  $\mu_{vol}$  estimation.

### A. Parameters Extraction

1) *Effective Length*: Fig. 1b shows the extraction of  $L_{eff}$  as described in Section II-A. One obtains  $\Delta L = 1.8$  nm. We verified that the obtained  $C_{gg,norm}/L_{eff}$  curves are well superimposed (not shown). Therefore, the influence of the  $C_{if}(V_{ov,min})$  term (see Section II-A) is considered negligible, validating the obtained  $\Delta L$  value in our case.

2) *Access Resistance*: Figure 1c shows the excellent linearity of  $R_m(L_{eff})$  curves, validating the use of the total resistance method. A significant bias dependence of  $R_{SD}$ , by  $\approx 50\%$  from  $V_{ov} = 0.3$  V to  $V_{ov,max} = 0.87$  V, as reported in [17], [18] is confirmed by our simulations.

3) *Inversion Charge*: Fig. 2b shows the linear interpolation of  $C_{gg}$  to extract the different coefficients described in

Section II-C. Results are shown on Fig. 2c. Eq.(4) is shown able to accurately model  $C_{gg}$  for both long and short transistors (with less than 2.5% error obtained for all gate lengths on the entire voltage range). The peak in  $B(V_{ov})$  observed below threshold is characteristic of the inner fringing capacitance [8], [17]. From these insights, we can evaluate the impact of the inner fringing capacitance on the extraction of  $Q_i$ . The inset in Fig. 2a indeed highlights the difference between  $C_{gg,norm}$  and  $A_1(V_{ov})L_{eff} \approx C_{inv}$  due to the inner fringing terms in Eq. (2). As shown in Fig. 2d, one then obtains a significant overestimation of  $Q_i$  when integrating  $C_{gg,norm}$  instead of  $A_1(V_{ov})L_{eff}$ , reaching 22% at  $V_{ov} = 0.3$  V for a  $L_G = 25$  nm transistor.

### B. Impact on Mobility

Fig. 2e,f, quantitatively show the impact of the different parasitic terms on the extracted  $\mu_{eff}$ . The simple split CV method (i.e. using  $C_{gg,norm}$  and a single  $R_{SD}$  value extracted at  $V_{ov,max}$ ) is shown unable to extract the implemented mobility behaviour in the short  $L_G = 25$  nm transistor, confirming the need for careful de-embedding of the parasitics. Previous articles have discussed the influence of the  $R_{SD}$  bias dependence on the low-field mobility  $\mu_0$  extracted with the Y-function method, notably showing it generates an erroneous mobility degradation with  $L_G$  [17], [18]. However, its impact on split CV extracted mobility values and their behaviour with  $V_{ov}$  has to the author's knowledge not been reported up to now. As Fig. 2f shows, it is particularly significant in the short transistor. The obtained unphysical behavior (monotone increase of  $\mu_{eff}$  with  $V_{ov}$ ) is corrected when including the significant bias dependence of  $R_{SD}$ . The remaining error (when using  $C_{gg,norm}$  but correcting for  $R_{SD}(V_{ov})$ ) is most pronounced in the  $L_G = 25$  nm transistor and at lower overdrive voltage. This is attributed to the inner fringing contribution in agreement with Fig. 2d. Using  $A_1L_{eff} \approx C_{inv}$  from the proposed modeling, we are then able to accurately extract the mobility behavior with  $V_{ov}$ . The extracted  $\mu_{eff}$  is within less than 10% error of the estimated TCAD value for all gate lengths in the  $0.3$  V  $< V_{ov} < 0.87$  V range, confirming the robustness of the proposed method.

## IV. VALIDATION ON 28 FDSOI MEASUREMENTS

Experimental data were finally used to verify the above insights. The measured devices are 28 FDSOI transistors from ST-Microelectronics [20] with  $L_G = 25, 35, 45, 60, 90$  and  $150$  nm and  $W = 120 \mu m$ . The gate capacitance (Fig. 3a) was extracted from the slope of  $Im(Y_{11}) = 2\pi f C_{gg}$  by linear interpolation against the frequency over  $f = [5, 10]$  GHz (Y parameters were obtained from open-pad de-embedded S parameters [21], bulk and source were kept to ground). Extraction of  $\Delta L$  is performed as described above and gives  $\Delta L = 6.5$  nm (inset in Fig. 3a). Figure 3b shows the extraction of  $R_{SD}(V_{ov})$ . The obtained bias dependence is in good agreement with the trend from Fig. 1d. The ability to model the total gate capacitance as  $C_{gg} = A(V_{ov})L_{eff} + B(V_{ov})$  is then verified in Fig. 3c. The gate capacitance is reconstructed from its linear interpolation coefficients with less than 2.5%

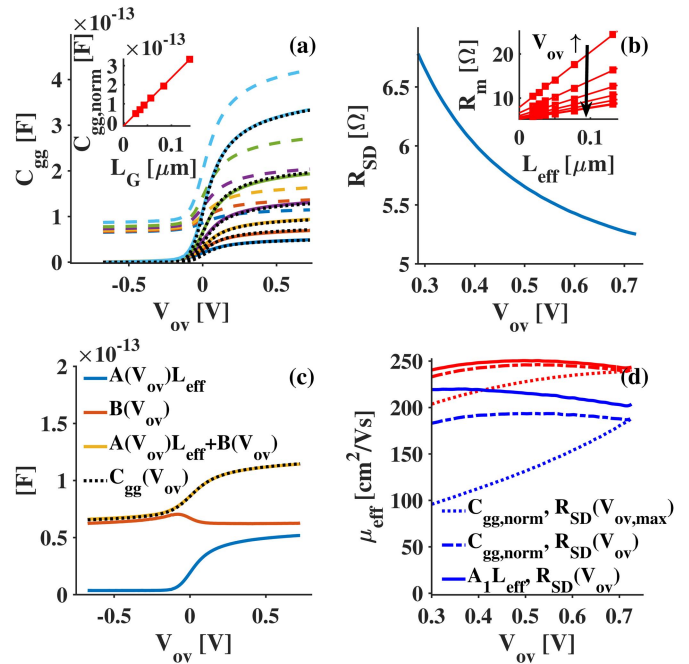


Fig. 3. (a)  $C_{gg}$  (dashed),  $C_{gg,norm}$  (full) and  $A_1L_{eff}$  (black dots) from 28 FDSOI measurements for  $L_G = [25, 35, 45, 60, 90, 150]$  nm (colors correspond to the legend in Fig. 2d) and  $V_{DS} = 50$  mV. The inset shows the extraction of  $L_{eff}$ . (b) Extraction of bias dependent  $R_{SD}$  with the total resistance method. Inset:  $R_m = \frac{V_{DS}}{I_D}$  versus  $L_{eff}$  for 6 equally spaced overdrive voltages in  $V_{ov} = [0.2, 0.7]$  V (c) Results of the linear interpolation of  $C_{gg}$  for a  $L_G = 25$  nm transistor (d) Extracted  $\mu_{eff}$  for the  $L_G = 150$  nm (red) and  $L_G = 25$  nm (blue) transistors.

relative error for all lengths and overdrive voltages. Finally, Fig. 3d compares the extracted  $\mu_{eff}$  when correcting for the different parasitics in both a long ( $L_G = 150$  nm) and short ( $L_G = 25$  nm) transistor. The results are in good agreement with the simulations presented in Fig. 2e,f.

## V. CONCLUSION

In this letter, we studied the application of the split CV method in 28 FDSOI transistors down to 25 nm gate length. The significant  $R_{SD}$  bias dependence related to the underlaps parasitic transistors was shown to have a major impact on the extracted mobility values and their evolution with the gate voltage. We also highlighted the impact of the inner fringing capacitance, whose contribution to the extracted charge was seen significant in short transistors. Correcting for these parasitics, the proposed method was shown able to extract  $\mu_{eff}$  within 10% error and to accurately reproduce its dependence on the gate voltage for transistors down to 25 nm gate length. The extraction only relies on direct device measurements, not requiring any pre-assumed expression of the inversion charge or mobility dependence on the gate voltage.

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